

Fourier transforms + frequency response functions

024 | 1/4

Fourier transforms can be used to characterize the steady state response of a system to a sinusoidal input. Let $u(t)$ be the sinusoidal input, $y(t)$ be the output, and $a_k, b_k \in \mathbb{R}$. A SISO system has the representation

$$\frac{d^m y}{dt^m} + a_{n-1} \frac{d^{n-1} y}{dt^{n-1}} + \dots + a_1 \frac{dy}{dt} + a_0 y = b_m \frac{d^m u}{dt^m} + \dots + b_1 \frac{du}{dt} + b_0 u.$$

We can take the Fourier transform of each side:

$$\mathcal{F}\left\{\frac{d^m y}{dt^m} + \dots + a_1 \frac{dy}{dt} + a_0 y\right\} = \mathcal{F}\left\{b_m \frac{d^m u}{dt^m} + \dots + b_1 \frac{du}{dt} + b_0 u\right\}$$

$$\mathcal{F}\left\{\frac{d^m y}{dt^m}\right\} + \dots + a_1 \mathcal{F}\left\{\frac{dy}{dt}\right\} + a_0 \mathcal{F}\{y\} = b_m \mathcal{F}\left\{\frac{d^m u}{dt^m}\right\} + \dots + b_1 \mathcal{F}\left\{\frac{du}{dt}\right\} + b_0 \mathcal{F}\{u\}$$

$$(j\omega)^m Y(j\omega) + \dots + a_1 (j\omega) Y(j\omega) + a_0 Y(j\omega) = b_m (j\omega)^m U(j\omega) + \dots + b_1 (j\omega) U(j\omega) + b_0 U(j\omega)$$

$$H(j\omega) = \frac{Y(j\omega)}{U(j\omega)} = \frac{b_m (j\omega)^m + \dots + j\omega b_1 + b_0}{(j\omega)^n + a_{n-1} (j\omega)^{n-1} + \dots + j\omega a_1 + a_0}. \quad (*)$$

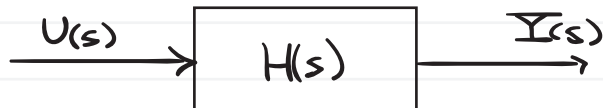
We call $H(j\omega)$ the **frequency response function**. It is a complex-valued function $H: \mathbb{R} \rightarrow \mathbb{C}$.

The following block diagrams are useful.

Frequency response function (Fourier):



Transfer function (Laplace):



System transfer operator (time-domain):



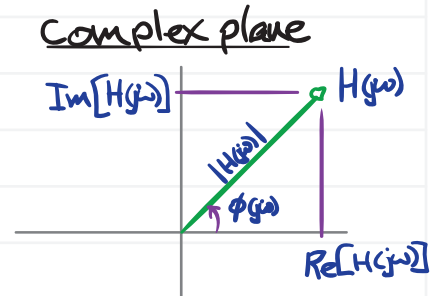
We can solve the SISO ODE by solving (*) for $Y(j\omega)$ and taking the inverse F.T.:

$$y(t) = \mathcal{F}^{-1}\{Y(j\omega)\} = \mathcal{F}^{-1}\{H(j\omega) U(j\omega)\}.$$

We usually consider the **magnitude** $|H(\omega)|$ and **phase** $\phi(\omega)$, which are defined as

$$|H(j\omega)| \equiv \left(\text{Re}[H(j\omega)]^2 + \text{Im}[H(j\omega)]^2 \right)^{1/2}$$

$$\phi(j\omega) \equiv \arctan \frac{\text{Im}[H(j\omega)]}{\text{Re}[H(j\omega)]}$$



Recall that we are only considering sinusoidal inputs. Assume

$$u(t) = A \sin(\omega t + \Psi).$$

We have seen that the steady-state response is

$$y(t) = A |H(\omega)| \sin(\omega t + \Psi + \phi(\omega)).$$

new amplitude

new phase

This was a very important result. It means that the steady-state response of a function to a sinusoidal input is sinusoidal and has a new amplitude and phase that depend on the frequency response function.

For this reason, the frequency response function's magnitude and phase give us significant insight into a system's s.s. response, and we like to plot them.

Bode plots: a review

Bode plots are how we typically plot $|H(j\omega)|$ and $\phi(j\omega)$.

First, recall that $X_{dB} \equiv 20 \log_{10} \frac{X_2}{X_1}$, where X_{dB} is a **decibel** representation of X_2 , X_2 is an amplitude, and X_1 is a reference amplitude.

We will be plotting $|H(j\omega)|$ on a log-dB plot. 024 3/4
 Since $|H(j\omega)|$ is already a ratio (of output to input), we can simply write

$$|H(j\omega)|_{dB} = 20 \log_{10} |H(j\omega)|$$

Example Draw a bode plot for the system described by the SISO ODE

$$\tau \dot{y} + y = k u$$

Solution

Fourier transform: $\tau(j\omega) \mathcal{F}\{y\} + \mathcal{F}\{y\} = k \mathcal{F}\{u\}$

$$\Rightarrow (j\omega\tau + 1) Y(j\omega) = k U(j\omega) \Rightarrow H(j\omega) \equiv \frac{Y(j\omega)}{U(j\omega)} = \frac{k}{1 + j\omega\tau}$$

Now we have the frequency response function. We can find its magnitude and phase after we find its real & imaginary components.

$$H(j\omega) = \frac{k}{1 + j\omega\tau} = \frac{k(1 - j\omega\tau)}{(1 - j\omega\tau)(1 + j\omega\tau)} = \frac{k(1 - j\omega\tau)}{1 + (\omega\tau)^2}$$

so the components are $\text{Re}[H(j\omega)] = \frac{k}{1 + (\omega\tau)^2}$ and $\text{Im}[H(j\omega)] = \frac{-\omega\tau k}{1 + (\omega\tau)^2}$

The magnitude and phase, then, are

$$\begin{aligned} |H(j\omega)| &= \left(\left(\frac{k}{1 + (\omega\tau)^2} \right)^2 + \left(\frac{-\omega\tau k}{1 + (\omega\tau)^2} \right)^2 \right)^{1/2} \\ &= \left(\frac{k^2 + \omega^2 \tau^2 k^2}{(1 + \omega^2 \tau^2)^2} \right)^{1/2} \\ &= \frac{k(1 + \omega^2 \tau^2)^{1/2}}{1 + \omega^2 \tau^2} \\ &= \frac{k}{\sqrt{1 + \omega^2 \tau^2}} \end{aligned}$$

$$\begin{aligned} \phi(j\omega) &= \text{atan} \frac{\frac{-\omega\tau k}{1 + (\omega\tau)^2}}{\frac{k}{1 + (\omega\tau)^2}} \\ &= \text{atan}(-\omega\tau) \\ &= -\text{atan}(\omega\tau) \end{aligned}$$

The plots are shown below.

