

## 10.2 freq.sin Sinusoidal input, frequency response

1 In this lecture, we explore the relationship—which turns out to be pretty chummy—between a system’s frequency response function  $H(j\omega)$  and its sinusoidal forced response.

2 Let’s build from the frequency response function  $H(j\omega)$  definition:

$$y(t) = \mathcal{F}^{-1}Y(\omega) \quad (1a)$$

$$= \mathcal{F}^{-1}(H(j\omega)U(\omega)). \quad (1b)$$

We take the input to be sinusoidal, with amplitude  $A \in \mathbb{R}$ , angular frequency  $\omega_0$ , and phase  $\psi$ :

$$u(t) = A \cos(\omega_0 t + \psi). \quad (2)$$

The Fourier transform of the input,  $U(\omega)$ , can be constructed via transform identities from [Table ft.1](#). This takes a little finagling. Let

$$p(t) = Aq(t), \quad (3a)$$

$$q(t) = r(t - t_0), \text{ and} \quad (3b)$$

$$r(t) = \cos \omega_0 t, \text{ where} \quad (3c)$$

$$t_0 = -\psi/\omega_0. \quad (3d)$$

The corresponding Fourier transforms, from [Table ft.1](#), are

$$P(\omega) = AQ(\omega), \quad (4a)$$

$$Q(\omega) = e^{-j\omega t_0}R(\omega), \text{ and} \quad (4b)$$

$$R(\omega) = \pi\delta(\omega - \omega_0) + \pi\delta(\omega + \omega_0). \quad (4c)$$

Putting these together,

(because  $\delta$ s)

3 And now we are ready to substitute into Eq. 1b; also applying the “linearity” property of the Fourier transform:

$$y(t) = A\pi (e^{j\psi}\mathcal{F}^{-1}(H(j\omega)\delta(\omega - \omega_0)) + e^{-j\psi}\mathcal{F}^{-1}(H(j\omega)\delta(\omega + \omega_0))). \quad (5)$$

The definition of the inverse Fourier transform gives

$$y(t) = \frac{A}{2} \left( e^{j\psi} \int_{-\infty}^{\infty} e^{j\omega t} H(j\omega) \delta(\omega - \omega_0) d\omega + e^{-j\psi} \int_{-\infty}^{\infty} e^{j\omega t} H(j\omega) \delta(\omega + \omega_0) d\omega \right). \quad (6)$$

Recognizing that  $\delta$  is an even distribution ( $\delta(t) = \delta(-t)$ ) and applying the sifting property of  $\delta$  allows us to evaluate each integral:

$$y(t) = \frac{A}{2} (e^{j\psi} e^{j\omega_0 t} H(j\omega_0) + e^{-j\psi} e^{-j\omega_0 t} H(-j\omega_0)). \quad (7)$$

Writing  $H$  in polar form,

(8)

The Fourier transform is **conjugate symmetric**—that is,  $F(-\omega) = F^*(\omega)$ —which allows us to further simplify:

$$y(t) = \frac{A|H(j\omega_0)|}{2} (e^{j(\omega_0 t + \psi)} e^{j\angle H(j\omega_0)} + e^{-j(\omega_0 t + \psi)} e^{-j\angle H(j\omega_0)}) \quad (9a)$$

$$= A|H(j\omega_0)| \frac{e^{j(\omega_0 t + \psi + \angle H(j\omega_0))} + e^{-j(\omega_0 t + \psi + \angle H(j\omega_0))}}{2}. \quad (9b)$$

Finally, Euler’s formula yields something that deserves a box.

**Equation 10 sinusoidal response in terms of  $H(j\omega)$**

4 This is a remarkable result. For an input sinusoid, a linear system has a forced response that

- is also a sinusoid,
- is at the same frequency as the input,
- differs only in amplitude and phase,
- differs in amplitude by a factor of  $|H(j\omega)|$ , and
- differs in phase by a shift of  $\angle H(j\omega)$ .

Now we see one of the key facets of the frequency response function: it governs how a sinusoid transforms through a system. And just think how powerful it will be once we combine it with the powerful principle of superposition and the mighty Fourier series representation of a function—as a “superposition” of sinusoids!